## Imperial College London

## Number Representations for Embedding Optimization Algorithms in Cyber-Physical Systems

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## http://cyberphysicalsystems.org

- "Cyber-Physical Systems (CPS) are integrations of computation, networking, and physical processes. Embedded computers and networks monitor and control the physical processes, with feedback loops where physical processes affect computations and vice versa...The technology builds on the older (but still very young) discipline of embedded systems, computers and software embedded in devices whose principle mission is not computation, such as cars, toys, medical devices, and scientific instruments."

| $\square$ | Advanced Search |
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| Authors » |
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| Publications " |
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| Keywords » |


| Academic > Top conferences in Real-Time \& Embedded Systems |  | 1-27 of 27 results |
| :---: | :---: | :---: |
| Computer Science $\quad$ Real-Time \& Embedded Systems | All Years |  |
| Conferences | Publications | H-Index - |
| RTSS - IEEE Real-Time Systems Symposium | 1160 | 85 |
| CDC - Conference on Decision and Control | 26379 | 83 |
| SenSys - Conference On Embedded Networked Sensor Systems | 635 | 69 |
| Hybrid Systems | 783 | 54 |
| CHES - Cryptographic Hardware and Embedded Systems | 428 | 53 |
| RTAS - IEEE Real Time Technology and Applications Symposium | 363 | 46 |
| ECRTS - Euromicro Conference on Real-Time Systems | 700 | 39 |
| CASES - Compilers, Architecture, and Synthesis for Embedded Systems | 407 | 34 |
| EMSOFT - International Workshop on Embedded Systems | 351 | 31 |
| ISORC - Object-Oriented Real-Time Distributed Computing | 770 | 29 |
| FTRTFT - Formal Techniques in Real-Time and Fault-Tolerant Systems | 218 | 27 |
| RTCSA - Real-Time Computing Systems and Applications | 815 | 26 |
| LCTES - Languages, Compilers, and Tools for Embedded Systems | 203 | 20 |

## CDC 2012 - Papers on "Networks"

| Communication networks | MoC02.3, ThA01.1, ThA01.2, ThA01.3, ThA01.4, ThA01.5, ThA01.6, ThA01.7, ThC01.4, ThC04.6, ThC08.3, ThC08.4, TuA01.6, TuA04.1, TuB01.2, TuB01.6, TuC02.3, TuC03.1, TuC04.2, WeA04.6, WeA17.4, WeB14.4 |
| :---: | :---: |
| Network analysis and control | MoA07.1, MoA07.6, MoB01.6, MoB11.1, ThA01.3, ThA01.4, ThA01.5, ThA01.7, ThA03.4, ThA06.2, ThA07.3, <br> ThA13.7, ThB03.3, ThB07.3, ThB14.2, ThB17.1, ThC01.3, ThC04.6, ThC07.1, ThC16.3, TuA01.2, TuA01.3, TuA01.4, <br> TuA03.2, TuA07.3, TuA10.3, TuA15.5, TuB01.2, TuB01.6, TuB02.1, TuB03.2, TuC01.2, TuC02.2, TuC02.5, TuC09.3, <br> WeA01.1, WeA01.3, WeA01.4, WeA01.5, WeA01.6, WeA01.7, WeA07.4, WeA14.4, WeA16.7, WeB01.1, WeB01.4, <br> WeB01.5, WeB01.6, WeB07.3, WeB09.5, WeB11.4, WeC01.4, WeC01.6 |
| Networked control systems | MoA01.1, MoA01.2, MoA01.3, MoA01.4, MoA01.5, MoA01.6, MoA01.7, MoA02.2, MoA02.3, MoA02.4, MoA02.7, MoA03.1, MoA03.6, MoA04.4, MoA07.1, MoA07.3, MoA07.7, MoA11.2, MoB01.1, MoB01.2, MoB01.3, MoB01.4, MoB01.5, MoB01.6, MoB02.5, MoB02.6, MoB03.3, MoB05.3, MoB11.4, MoB17.4, MoC01.1, MoC01.2, MoC01.3, MoC01.4, MoC01.5, MoC01.6, MoC02.6, ThA01.1, ThA01.5, ThA03.6, ThA05.7, ThA06.5, ThA16.1, ThB01.1, ThB01.2, ThB01.3, ThB01.4, ThB01.5, ThB03.2, ThB03.5, ThB03.6, ThB09.2, ThB10.6, ThC01.1, ThC01.2, ThC01.3, ThC01.4, ThC01.5, ThC01.6, ThC07.2, ThC07.3, ThC08.1, ThC10.6, TuA01.1, TuA01.2, TuA01.3, TuA01.4, TuA01.5, TuA01.6, TuA01.7, TuA02.2, TuA03.2, TuA06.4, TuA06.6, TuA10.4, TuA12.6, TuA17.4, TuB01.5, TuB02.3, TuB02.6, TuB17.1, TuC01.1, TuC01.2, TuC01.3, TuC01.4, TuC01.5, TuC02.3, TuC02.4, TuC04.3, TuC11.1, TuC17.5, WeA01.2, WeA03.5, WeA04.6, WeA06.2, WeA07.3, WeB01.2, WeB01.5, WeB03.1, WeB03.3, WeB07.2, WeB07.4, WeB07.6, WeB08.5, WeC03.3, WeC03.4, WeC14.2, WeC16.4 |
| Sensor networks | MoA01.4, MoA02.1, MoA02.3, MoB02.1, MoB02.2, MoB02.3, MoB02.4, MoB02.5, MoB02.6, MoB03.1, MoB03.2, MoB13.2, MoC02.1, MoC02.2, MoC02.3, MoC02.4, MoC02.5, MoC02.6, ThC08.2, ThC08.3, TuA02.1, TuA02.6, TuB02.1, TuB02.2, TuB02.3, TuB02.4, TuB02.5, TuB02.6, TuB04.5, TuC02.1, TuC02.2, TuC02.3, TuC02.4, TuC02.5, TuC02.6, TuC04.2, TuC04.3, TuC06.4, TuC14.1, TuC14.5, WeA03.1, WeA05.3, WeB07.1, WeB14.5, WeC01.2, WeC01.4, WeC07.4, WeC07.5 |
| Transportation networks | MoA07.2, MoA07.5, MoA12.1, MoA12.6, ThB12.6, ThB16.1, ThB16.2, ThC07.1, WeC14.1 |

## CDC 2012 - Papers on "Computation"

 MoC16.3, ThA05.1, ThA12.3, ThA14.6, ThA17.7, ThB06.4, ThC06.4, ThC13.2, ThC14.1, ThC14.5, TuA02.5, TuA06.5, TuB09.1, TuB09.5, TuC05.3, TuC05.6, TuC06.2, TuC15.3, WeA01.6, WeA04.4, WeA10.7, WeA13.3, WeA17.2, WeB10.3, WeC09.2, WeC10.2, WeC14.4, WeC17.5
## CDC 2012 - Papers on "Embedded Systems"

## CDC 2012 - Papers on "Real-time Systems"

## Embedded Optimization (Optimal Control / <br> Estimation / DSP) for Cyber-Physical Systems

Applications for Embedded Optimization (Optimal Control / Estimation / DSP)


C우다튼튼


## Computing and Cyber-Physical Systems

"All too often, today's students use laptop [or desktop] computers to perform their computing, which shields them from dealing with any of the physical constraints they will face in the real world. This approach is akin to trying to learn skiing while standing comfortably in the après ski lounge."

Wolf, Cyber-Physical Systems, Computer, 2009.

## Challenges for Cyber-Physical Systems

- Cost
- Energy
- Speed
- Reliability

- Predictability / real-time (fast is not equal to real-time)

Number representation (e.g. fixed/floating-point, \#bits) has a major impact on the design

## Size is Very Important in Microprocessor Design



Cost per die $=f\left(\right.$ area $\left.^{x}\right), \quad \boldsymbol{x} \in[2,4]$

## Computational Resources for an Adder

Xilinx Virtex-7 XT 1140 FPGA:

| Number <br> representation | Registers/Flip- <br> Flops (FFs) | Latency/delay <br> (clock cycles) |
| :---: | :---: | :---: |
| double floating-point <br> 52-bit mantissa | 1046 | 14 |
| single floating-point <br> 23-bit mantissa | 557 | 11 |
| fixed-point <br> 53 bits | 53 | 1 |
| fixed-point <br> 24 bits | 24 | 1 |

Cheap and low power processors often only have fixed-point

## Dynamic Optimization

$$
\begin{gathered}
\min _{x(\cdot), u(\cdot), p} J(y(\cdot), x(\cdot), u(\cdot), p) \\
F(y(t), \dot{x}(t), x(t), u(t), p, t)=0, \quad \forall t \in\left[t_{0}, t_{f}\right) \\
G(y(t), \dot{x}(t), x(t), u(t), p, t) \leq 0,
\end{gathered} \forall t \in\left[t_{0}, t_{f}\right) \text {. }
$$

Discretized and approximated by finite-dimensional NLP:

$$
\begin{gathered}
\min _{\theta} V(\theta) \\
f(\theta)=0 \\
g(\theta) \leq 0 \\
\theta \in \mathbb{R}^{n}, f: \mathbb{R}^{n} \rightarrow \mathbb{R}^{m}, g: \mathbb{R}^{n} \rightarrow \mathbb{R}^{p}
\end{gathered}
$$

Fixed-Point Arithmetic

Floating-Point Arithmetic

## round-off <br> 1100000110011110001. ..eror

round-off
010101000100111111 ...eron
time
round-off
00100100110011000 ...ror

## Fixed-Point Arithmetic



Challenges for Fixed-Point Arithmetic

- Number of bits for integer and fractional part?
- Determine worst-case peak values
- Optimization algorithms are nonlinear and recursive
- Search direction most computationally critical part:

$$
\mathrm{A} \xi=\mathrm{b}
$$

- Iterative linear solvers preferred: CG, MINRES, GMRES


## Lanczos Algorithm (Kernel of CG/MINRES)

$$
\mathrm{Q}_{i}^{T} \mathrm{AQ}_{i}=\mathrm{T}_{i}:=\left[\begin{array}{cccc}
\alpha_{1} & \beta_{1} & & 0 \\
\beta_{1} & \alpha_{2} & \ddots & \\
& \ddots & \ddots & \beta_{i-1} \\
0 & & \beta_{i-1} & \alpha_{i}
\end{array}\right]
$$

Given $q_{1}$ such that $\left\|q_{1}\right\|_{2}=1$ and an initial value $\beta_{0}:=1$ for $i=1$ to $i_{\max }$ do

1. $q_{i} \leftarrow \frac{q_{i}}{\beta_{i-1}}$
2. $z_{i} \leftarrow \mathrm{~A} q_{i}$
3. $\alpha_{i} \leftarrow q_{i}^{T} z_{i}$
4. $q_{i+1} \leftarrow z_{i}-\alpha q_{i}-\beta_{i-1} q_{i-1}$
5. $\beta_{i} \leftarrow\left\|q_{i+1}\right\|_{2}$
end for

## Lanczos Algorithm (Kernel of CG/MINRES)

$$
\mathrm{Q}_{i}^{T} \mathrm{AQ}_{i}=\mathrm{T}_{i}:=\left[\begin{array}{cccc}
\alpha_{1} & \beta_{1} & & 0 \\
\beta_{1} & \alpha_{2} & \ddots & \\
& \ddots & \ddots & \beta_{i-1} \\
0 & & \beta_{i-1} & \alpha_{i}
\end{array}\right]
$$

Given $q_{1}$ such that $\left\|q_{1}\right\|_{2}=1$ and an initial value $\beta_{0}:=1$ for $i=1$ to $i_{\max }$ do

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2. $z_{i} \leftarrow \mathrm{~A} q_{i}$
3. $\alpha_{i} \leftarrow q_{i}^{T} z_{i}$
4. $q_{i+1} \leftarrow z_{i}-\alpha q_{i}-\beta_{i-1} q_{i-1}$
5. $\beta_{i} \leftarrow\left\|q_{i+1}\right\|_{2}$
end for

## Evolution of Variables in Primal-dual Interior Point



Optimal control of a Boeing 747

## On-line Diagonal Preconditioner / Scaler

$$
\begin{gathered}
\mathrm{A} \xi=\mathrm{b}, \quad \mathrm{~A}=\mathrm{A}^{\prime} \\
\mathrm{S}_{k k}:=\sum_{j=1}^{\mathrm{N}}\left|\mathrm{~A}_{k j}\right| \quad(1 \text {-norm of row } k) \\
\mathrm{S}^{-\frac{1}{2}} \mathrm{AS}^{-\frac{1}{2}} \psi=\mathrm{S}^{-\frac{1}{2}} \mathrm{~b} \Leftrightarrow \widehat{\mathrm{~A}} \psi=\widehat{\mathrm{b}} \Rightarrow \rho(\widehat{\mathrm{~A}}) \leq 1 \\
\xi=\mathrm{S}^{-\frac{1}{2}} \psi
\end{gathered}
$$

Theorem (Avoiding overflow in fixed-point) All variables in Lanczos algorithm are between -2 and 2

Proof: Proc. IEEE Conference on Decision and Control 2012

## Evolution of Variables in Primal-dual Interior Point



Optimal control of a Boeing 747

## Trade-offs on an FPGA (same accuracy)



Xilinx Virtex-7 XT 1140 with matrices from the optimal control of a Boeing 747

Low-Precision Arithmetic

## Optimal Control in Low Precision Arithmetic




Mass-spring system with 3 masses ( 6 states) and 2 inputs, sample period $=10 \mathrm{~ms}$

## Sampled-data Representation in Shift Form

$$
\dot{x}(t)=A_{c} x(t)+B_{c} u(t)
$$

Sample period $h$ and piecewise constant input (ZOH):

$$
u(t)=u(k h)=: u_{k}, \quad \forall t \in[k h, k h+h)
$$

Exact solution/discrete-time model to compute $x_{k}:=x(k h)$

$$
\begin{gathered}
x_{k+1}=A_{s} x_{k}+B_{s} u_{k} \\
A_{s}:=e^{A_{c} h}=I+A_{c} h+\frac{\left(A_{c} h\right)^{2}}{2!}+\frac{\left(A_{c} h\right)^{3}}{3!}+\ldots \\
\lim _{\left\|A_{c} h\right\| \rightarrow 0} A_{s}=I, \quad \lim _{\left\|A_{c} h\right\| \rightarrow 0} B_{s}=0
\end{gathered}
$$

## Sampled-data Representation in Shift Form

$$
\dot{x}(t)=A_{c} x(t)+B_{c} u(t)
$$

Sample period $h$ and piecewise constant input (ZOH):

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\end{gathered}
$$

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\dot{x}(t)=A_{c} x(t)+B_{c} u(t)
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Sample period $h$ and piecewise constant input (ZOH):

$$
u(t)=u(k h)=: u_{k}, \quad \forall t \in[k h, k h+h)
$$

Exact solution/discrete-time model to compute $x_{k}:=x(k h)$

$$
\begin{gathered}
\left(\left(x_{k+1}-x_{k}\right) / h=\left(A_{s} x_{k}+B_{s} u_{k}-x_{k}\right) / h\right. \\
A_{s}:=e^{A_{c} h}=I+A_{c} h+\frac{\left(A_{c} h\right)^{2}}{2!}+\frac{\left(A_{c} h\right)^{3}}{3!}+\ldots \\
\lim _{\left\|A_{c} h\right\| \rightarrow 0} A_{s}=I, \quad \lim _{\left\|A_{c} h\right\| \rightarrow 0} B_{s}=0
\end{gathered}
$$

## Sampled-data Representation in Delta Form

Middleton and Goodwin (IEEE TAC, 1986):

$$
\left(x_{k+1}-x_{k}\right) / h=\left(A_{s} x_{k}+B_{s} u_{k}-x_{k}\right) / h
$$

## Sampled-data Representation in Delta Form

Middleton and Goodwin (IEEE TAC, 1986):

$$
\frac{x_{k+1}-x_{k}}{h}=\frac{\left(A_{s}-I\right)}{h} x_{k}+\frac{B_{s}}{h} u_{k}
$$

## Sampled-data Representation in Delta Form

Middleton and Goodwin (IEEE TAC, 1986):

$$
\frac{x_{k+1}-x_{k}}{h}=A_{\delta} x_{k}+B_{\delta} u_{k}
$$

## Sampled-data Representation in Delta Form

Middleton and Goodwin (IEEE TAC, 1986):

$$
\begin{gathered}
\frac{x_{k+1}-x_{k}}{h}=A_{\delta} x_{k}+B_{\delta} u_{k} \\
A_{\delta}=A_{c}+\frac{A_{c}^{2} h}{2!}+\frac{A_{c}^{3} h^{2}}{3!}+\ldots \\
\lim _{\left\|A_{c} h\right\| \rightarrow 0} A_{\delta}=A_{c}, \quad \lim _{\left\|A_{c} h\right\| \rightarrow 0} B_{\delta}=B_{c}
\end{gathered}
$$

Equivalent to shift form in infinite precision arithmetic Different from shift form in finite precision arithmetic

## Optimization Problem Using Shift Form

$$
\left.\begin{array}{c}
\min _{\theta} \sum_{k=0}^{N-1} \ell_{k}\left(x_{k}, u_{k}\right) \\
\theta:=\left[\begin{array}{llllll}
u_{0}^{\prime} & x_{1}^{\prime} & u_{1}^{\prime} & x_{2}^{\prime} & \cdots & u_{N-1}^{\prime}
\end{array} x_{N}^{\prime}\right.
\end{array}\right]^{\prime} .
$$

subject to

$$
\begin{array}{cc}
x_{0}=\hat{x}, & \\
x_{k+1}=A_{s} x_{k}+B_{s} u_{k}, & \forall k \in\{0,1, \ldots, N-1\} \\
J x_{k}+E u_{k} \leq d, & \forall k \in\{0,1, \ldots, N-1\}
\end{array}
$$

## Optimization Problem Using Shift Form

$$
\left.\begin{array}{c}
\min _{\theta} \sum_{k=0}^{N-1} \ell_{k}\left(x_{k}, u_{k}\right) \\
\theta:=\left[\begin{array}{llllll}
u_{0}^{\prime} & x_{1}^{\prime} & u_{1}^{\prime} & x_{2}^{\prime} & \cdots & u_{N-1}^{\prime}
\end{array} x_{N}^{\prime}\right.
\end{array}\right]^{\prime} .
$$

subject to

$$
\begin{array}{cc}
x_{0}=\hat{x}, & \\
x_{k+1}=A_{s} x_{k}+B_{s} u_{k}, & \forall k \in\{0,1, \ldots, N-1\} \\
J x_{k}+E u_{k} \leq d, & \forall k \in\{0,1, \ldots, N-1\}
\end{array}
$$

$$
\delta_{k}:=\frac{x_{k+1}-x_{k}}{h}=A_{\delta} x_{k}+B_{\delta} u_{k}
$$

## Optimization Problem Using Delta Form

$$
\begin{gathered}
\min _{\theta} \sum_{k=0}^{N-1} \ell_{k}\left(x_{k}, u_{k}\right) \\
\theta:=\left[\begin{array}{llllll}
u_{0}^{\prime} & \delta_{0}^{\prime} & x_{1}^{\prime} & u_{1}^{\prime} & \delta_{1}^{\prime} & x_{2}^{\prime} \\
\cdots & u_{N-1}^{\prime} & \delta_{N-1}^{\prime} & x_{N}^{\prime}
\end{array}\right]^{\prime}
\end{gathered}
$$

subject to

$$
\begin{gathered}
x_{0}=\hat{x}, \\
\begin{array}{r}
\delta_{k}=A_{\delta} x_{k}+B_{\delta} u_{k}, \\
x_{k+1}=x_{k}+h \delta_{k},
\end{array} \quad \forall k \in\{0,1, \ldots, N-1\} \\
J x_{k}+E u_{k} \leq d, \quad \forall k \in\{0,1, \ldots, N-1\} \\
\delta_{k}:=\frac{x_{k+1}-x_{k}}{h}=A_{\delta} x_{k}+B_{\delta} u_{k}
\end{gathered}
$$

## Optimization Problem Using Delta Form

$$
\begin{gathered}
\min _{\theta} \sum_{k=0}^{N-1} \ell_{k}\left(x_{k}, u_{k}\right) \\
\theta:=\left[\begin{array}{llllll}
u_{0}^{\prime} & \delta_{0}^{\prime} & x_{1}^{\prime} & u_{1}^{\prime} & \delta_{1}^{\prime} & x_{2}^{\prime} \\
\cdots & u_{N-1}^{\prime} & \delta_{N-1}^{\prime} & x_{N}^{\prime}
\end{array}\right]^{\prime}
\end{gathered}
$$

subject to

$$
\begin{gathered}
x_{0}=\hat{x}, \\
\begin{array}{r}
\delta_{k}=A_{\delta} x_{k}+B_{\delta} u_{k}, \\
x_{k+1}=x_{k}+h \delta_{k},
\end{array} \\
\begin{aligned}
& J x_{k}+E u_{k} \leq d, \forall k \in\{0,1, \ldots, N-1\} \\
& \forall k \in\{0,1, \ldots, N-1\} \\
&\hline 0,1, \ldots, N-1\}
\end{aligned}
\end{gathered}
$$

## Solving the Optimization Problem

- Solve linearized KKT system (Rao, Wright, Rawlings; JOTA, 1998):

$$
\mathrm{A} \xi=\mathrm{b}
$$

- Interleave search direction variables:

$$
\xi:=\left[\begin{array}{lllllll}
\Delta u_{0}^{\prime} & \Delta \gamma_{0}^{\prime} & \Delta \delta_{0}^{\prime} & \Delta \lambda_{1}^{\prime} & \Delta x_{1}^{\prime} & \cdots & \Delta x_{N}^{\prime}
\end{array}\right]^{\prime}
$$

- Block elimination results in Riccati recursions:

$$
\begin{array}{r}
P_{k-1}:=Q_{k-1}+P_{k}+h^{2} A_{\delta}^{\prime} P_{k} A_{\delta}+h A_{\delta}^{\prime} P_{k}+h P_{k} A_{\delta} \\
-\left(M_{k-1}+h^{2} A_{\delta}^{\prime} P_{k} B_{\delta}+h P_{k} B_{\delta}\right)\left(R_{k-1}+h^{2} B_{\delta}^{\prime} P_{k} B_{\delta}\right)^{-1} \\
\left(M_{k-1}^{\prime}+h^{2} B_{\delta}^{\prime} P_{k} A_{\delta}+h B_{\delta}^{\prime} P_{k}\right)
\end{array}
$$

## Data Dependencies in Riccati Recursion



Same amount of multipliers, adders and computational delay for a custom circuit, e.g. FPGA

## Optimal Control in Low Precision Arithmetic




Mass-spring system with 3 masses ( 6 states) and 2 inputs, sample period $=10 \mathrm{~ms}$

## Conclusions

- Number representation major factor that determines cost, energy, computational delay and accuracy
- Fixed-point: Precondition to get tight analytical bounds on variables in Lanczos algorithm to avoid overflow
- Low precision: Sampled-data model and optimization method crucial to successful implementation
- Co-design algorithm and hardware to use "just the right amount" of computational resources


## Open Research Questions

- Other sampled-data and number representations?
- Nonlinear systems?
- Which algorithms map easily to low precision, fixed-point or other number representations?
- A priori guarantees on accuracy, closed-loop stability, robustness and performance?
- Need control + optimization + numerics + computing

